

Simplify The Calculation Of Microstrip Dimensions

Microwaves and RF

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The classic equations used to predict the dimensions of microstrip lines can be greatly simplified without a significant sacrifice in the accuracy of those predictions.

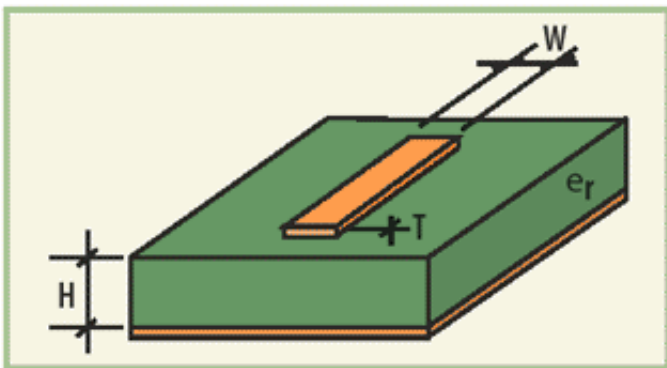
Circuit traces on microstrip printed circuit boards (PCBs) form the interconnections between components in addition to the components themselves. These controlled-impedance lines are typically designed with a characteristic impedance of 50 Ω for RF/microwave circuits or 75 Ω for cable television (CATV). The equations commonly associated with calculating the geometry of microstrip lines based on a desired impedance are quite complex and not easy to handle. Fortunately, a set of simplified equations has been developed based on well-known publications from IPC (www.ipg.org).

In these simplified equations, the dimensions for thickness, T, must be expressed in microns (m), although the values for height, H, and width, W, can have any measurement unit congruent with one another. The plots shown throughout the article assume a copper thickness of 35 m with the errors are calculated as:

$$Error = 100 \cdot \frac{Z_{calc} - Z_{theor}}{Z_{theor}} \quad (A)$$

where:

Z_{cal} = the calculated impedance and Z_{theor} = the expected value (i.e., 50 or 75 Ω)



1. This simple cross-sectional diagram shows the basic dimensional parameters of a microstrip line on a printed-circuit board with relative dielectric constant ϵ_r .

[Figure 1](#) shows a cutaway view of a typical microstrip circuit with the three dimensions and the relative dielectric constant (ϵ_r). In the IPC's publication IPC-2141, the impedance of a microstrip line is given by:

$$Z_0 = \frac{87}{\sqrt{\epsilon_r + 1.41}} \cdot \ln\left(\frac{5.98 \cdot H}{0.8 \cdot W + T}\right) \quad (B)$$

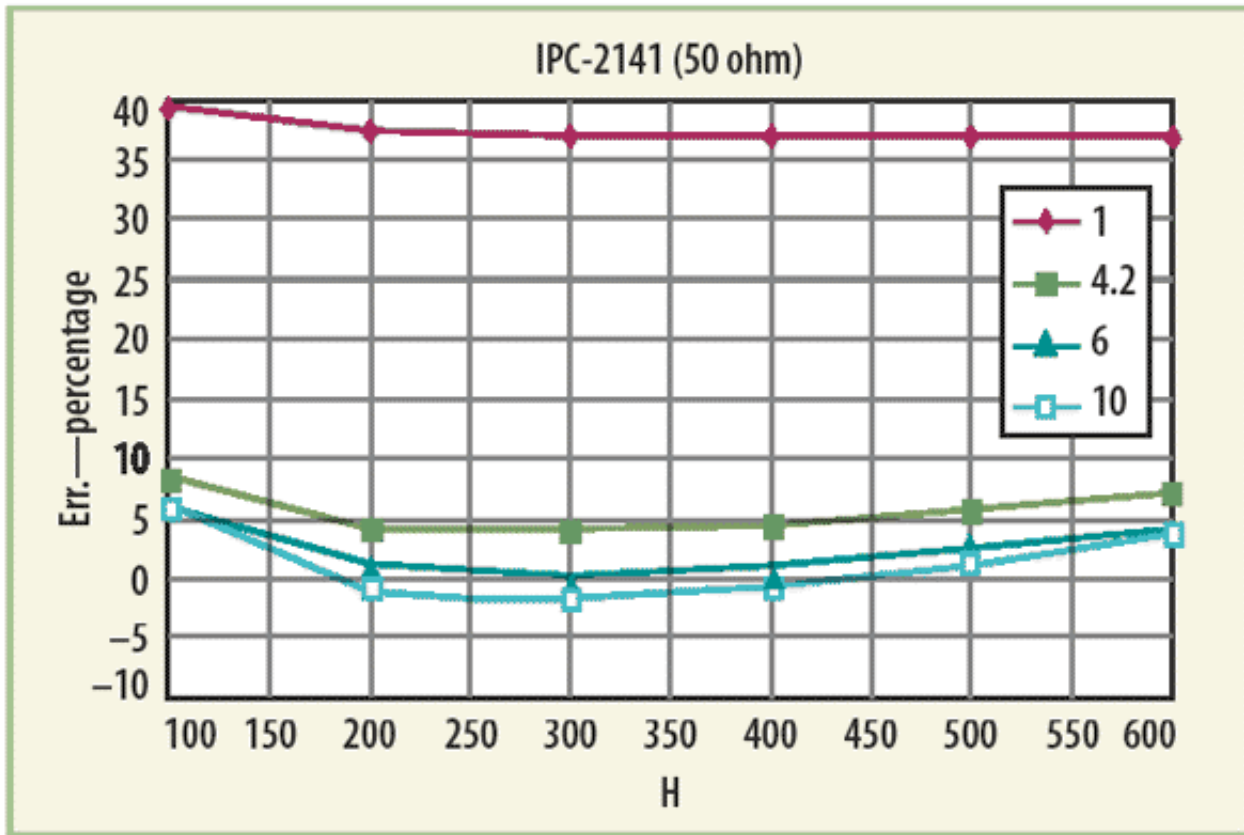
This relationship can be rearranged to show width, W, explicitly as:

$$W = \frac{7.48 \cdot H}{e^{\frac{Z_0}{87} \sqrt{(\epsilon_r + 1.41)}}} - 1.25 \cdot T \quad (C)$$

When W is determined for characteristic impedance (Z_0 of $50\ \Omega$ or $75\ \Omega$ the calculations can be performed by means of:

$$W = \begin{cases} \frac{7.48 \cdot H}{e^{\sqrt{0.33 \cdot (\epsilon_r + 1.41)}} - 1.25 \cdot T} & Z_0 = 50\ \Omega \\ \frac{7.48 \cdot H}{e^{\sqrt{0.74 \cdot (\epsilon_r + 1.41)}} - 1.25 \cdot T} & Z_0 = 75\ \Omega \end{cases} \quad (1)$$

Plots of errors based on circuit-board materials with several different dielectric constants were determined by using the previous equations. The results are shown in [Fig. 2](#).



2. The expected errors for predicting W are shown as a function of H for the IPC-2141 50- Ω case.

To simplify the microstrip dimensional formula, the exponential term can be linearized. This linear series expansion leads to:

$$e^{\sqrt{x}} = e^{\sqrt{x_0}} + \frac{1}{2 \cdot \sqrt{x_0}} \cdot e^{\sqrt{x_0}} \cdot (x - x_0) \quad (D)$$

Assuming a range of dielectric constants (ϵ_r) from 3 to 10 as in the original formula, the value of $x_0 = 2.6$ for a

characteristic impedance (Z_0) of 50Ω and the value of $x_0 = 5.9$ for a characteristic impedance of 75Ω so that:

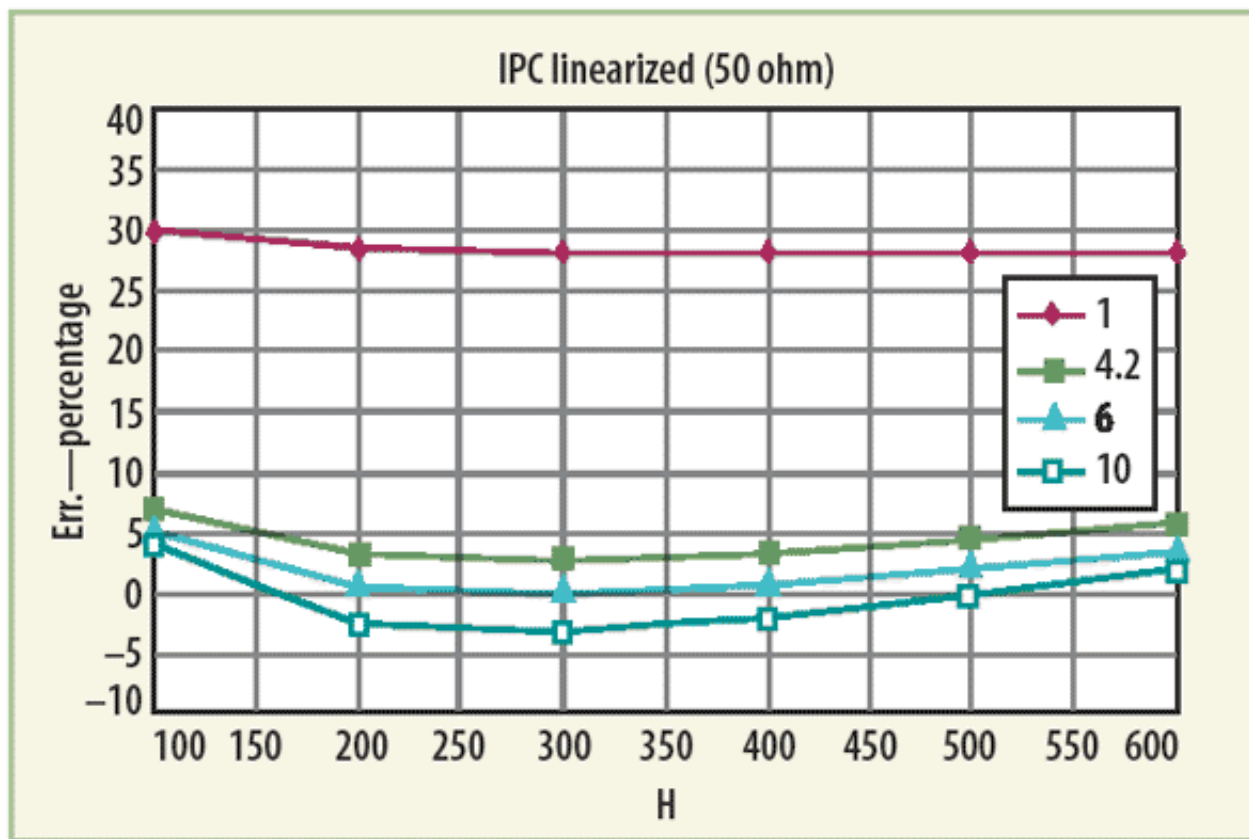
$$W_{lin} = \begin{cases} \frac{7.48 \cdot H}{1.56 \cdot 0.33 \cdot (\epsilon_r + 1.41) + 0.97 \epsilon_r} - 1.25 \cdot T & Z_0 = 50 \Omega \\ \frac{7.48 \cdot H}{2.34 \cdot 0.74 \cdot (\epsilon_r + 1.41) - 2.43 \epsilon_r} - 1.25 \cdot T & Z_0 = 75 \Omega \end{cases} \quad (E)$$

Page Title

This can be further simplified to:

$$W_{lin} = \begin{cases} \frac{14.67 \cdot H}{\epsilon_r + 3.31} - 1.25 \cdot T & Z_0 = 50 \Omega \\ \frac{4.32 \cdot H}{\epsilon_r} - 1.25 \cdot T & Z_0 = 75 \Omega \end{cases} \quad (F)$$

The error curves are shown in [Fig.3](#) using the linearized relationships and assuming microstrip lines with characteristic impedance of 50Ω .



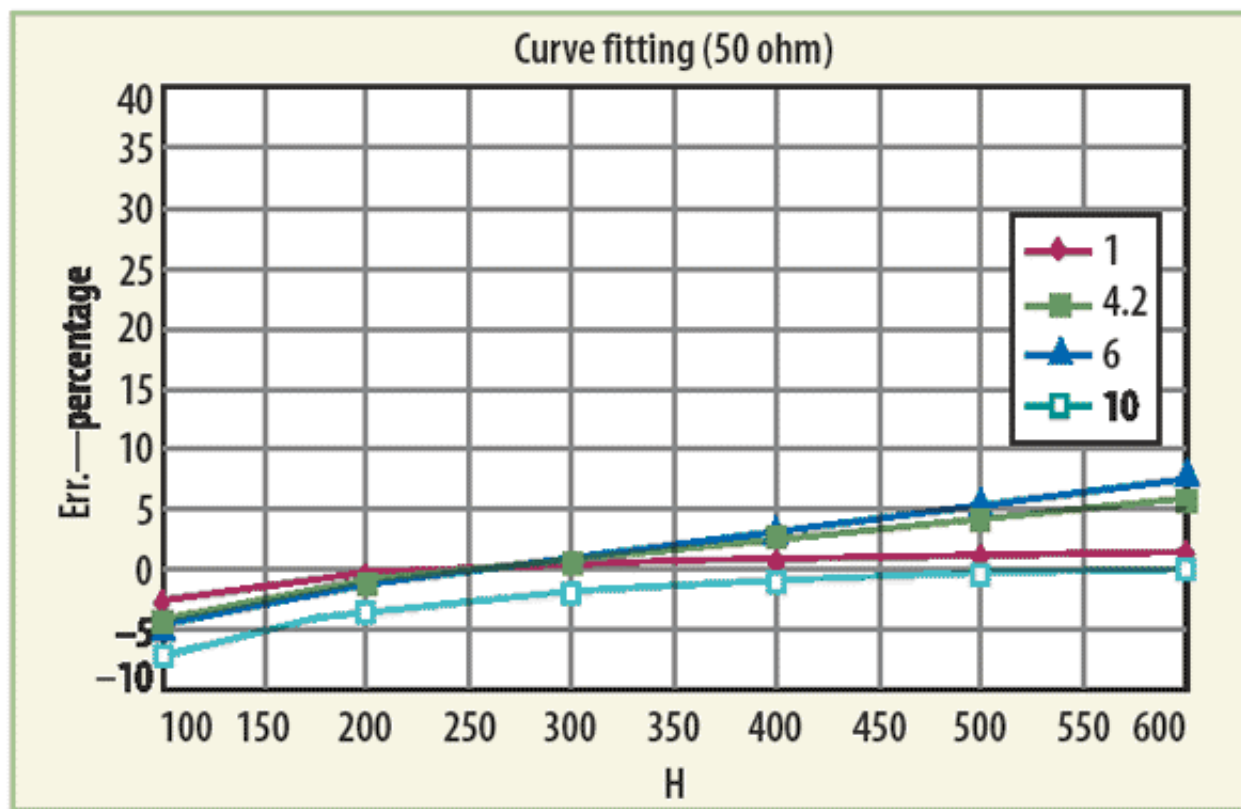
3. The expected errors for predicting W are shown as a function of H for the IPC-2141 $50\text{-}\Omega$ linearized case.

Due to the good results obtained by the linearized formula, the same model was used to fit the values from an

EM field solver. This approach led to a surprisingly simple equation for the case of a characteristic impedance (Z_0) of 50 Ω :

$$W = \frac{10 \cdot H}{1 + \epsilon_r + \frac{T}{300}} \quad (G)$$

The curves shown in Fig.4 are the result of curve-fitting the error curves that were generated under the same conditions as the original IPC formula and the linearized IPC-based formula.



4. These curves are the results of curve-fitting standard 50- Ω IPC predictions and the values given by the simplified width predictions.

It is possible to make a further simplification/approximation by disregarding the value of the microstrip line thickness, T:

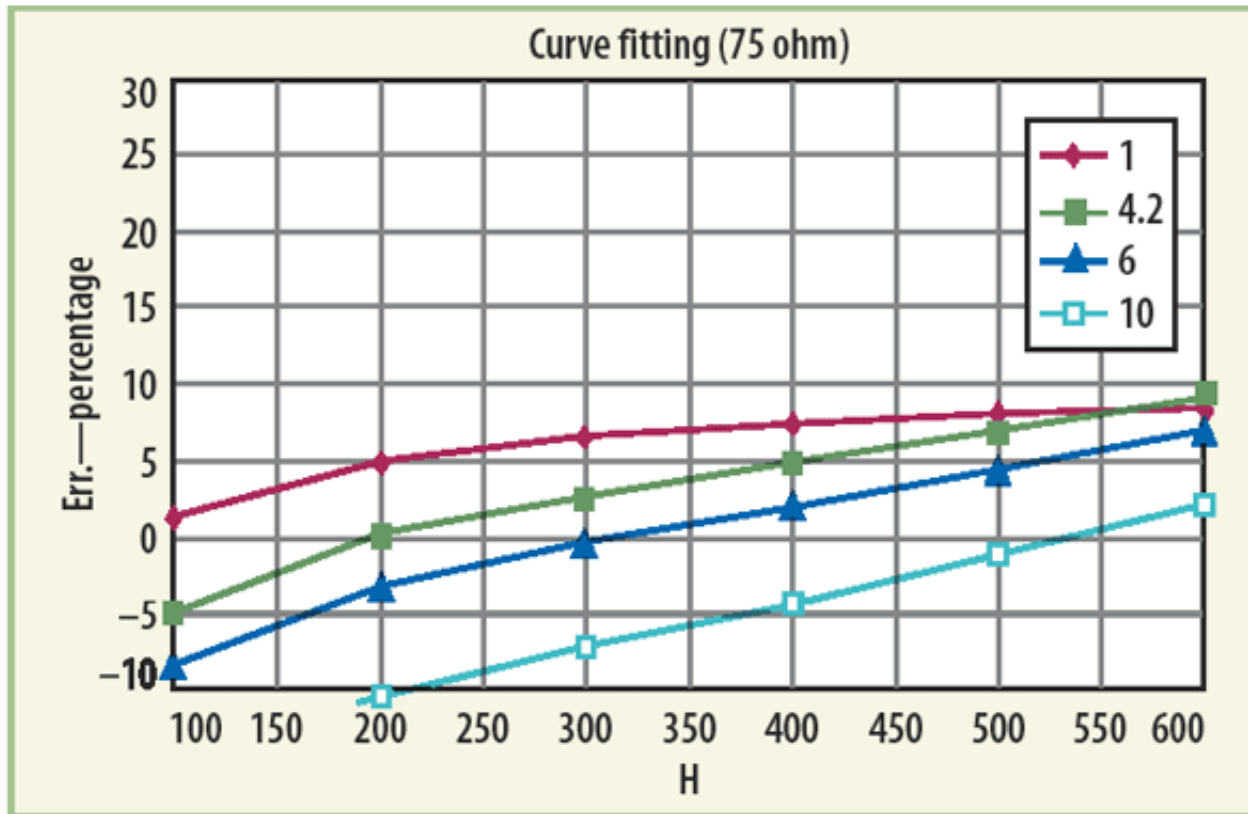
$$W_{50\Omega} = \frac{10 \cdot H}{1 + \epsilon_r} \quad (H)$$

This last equation confirms the rule of thumb that the ratio of the width to height of a 50 Ω line, W/H, on FR4 circuit-board material, is approximately 2. In fact, for common FR4, the relative dielectric constant (ϵ_r) is approximately 4.

Shifting to the curve for a characteristic impedance (Z_0) of 75 Ω the following equation results:

$$W = \frac{3.85 \cdot H}{0.6 + \epsilon_r + \frac{T}{300}} \quad (I)$$

with the same-curve-fitting procedure used to generate the curves shown in Fig.5.



5. These curves are the results of curve-fitting standard 75-Ω IPC predictions and the values given by the simplified width predictions.

As in the case with the 50 Ω characteristic impedance, if the microstrip thickness parameter, T, is neglected, the resulting simplified equation is:

$$W_{75\Omega} = \frac{3.85 \cdot H}{0.6 + \epsilon_r} \quad (J)$$

These simple solutions can be used for a wide range of RF and microwave microstrip calculations at 50 and 75 Ω.

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